

# Low parasitic inductance multi-chip SiC devices packaging technology

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## Keywords

« Wide bandgap devices », « High frequency power converter », « Silicon Carbide (SiC) », « Packaging », « high voltage power converters », « JFET » « Wind energy »

## Abstract

This paper presents a novel packaging structure which employs stacked substrate and flexible printed circuit board (PCB) to obtain very low parasitic inductance and hence feature high switching speed SiC power devices. A half-bridge module aimed at blocking voltage up to 2.5kV has been designed to accommodate 8 SiC JFETs and 4 SiC diodes. Electromagnetic simulation results reveal extremely low inductance values of the major loops. Then the prototyping of the designed package including the assembly process, all the electrical test to evaluate the electrical performance are presented.

## Introduction

Power semiconductor devices have largely improved since they were first introduced more than 50 years ago. The recent availability of silicon carbide (SiC) devices has pushed the boundaries even further in terms of power density, conversion efficiency, switching speed or thermal capability. In fact, they are enabling components mainly if higher switching frequencies in power electronics are considered. However, this trend imposes new challenges towards the package of these devices where stray parasitic inductances become crucial concern specially to avoid the high spikes due to high  $di/dt$  and  $dV/dt$  under high switching frequency. Hence, new packaging solution is needed to reduce the stray parasitic inductance for addressing this concern.

On the other hand, the most common technology of a die-level interconnect is wire bonding. Maturity, flexibility and low cost are main factors, which explain the use of the wire bonding technology. However, it has electrical limitation (such as parasitic inductance) and thermo-mechanical limitations such as thermal stress leading to the fatigue and eventually a failure. Alternative interconnection technologies have been developed to overcome these shortcomings, allowing a three-dimensional (3D) packaging of power modules in a compact stacked layer structure [1]–[7]. The technology used in the reported works mainly uses a conventional layout of DBC substrates on both sides of the devices to allow a double-side

cooling. It should also be noted that the majority of the work reported was applied to low-voltage devices and applications.

In this paper, a new design of a power module is presented, with very low parasitic inductance specifications. The presented design is targeting the relatively high voltage SiC devices and it's based on an optimization of the conductors' geometry, having the input and output conductors to be parallel with narrow separation, and hence, cancelling by that the electromagnetic field that determines the inductance. The concept was recently reported [8]. In this paper, the assembling process is presented in details and the assembled prototype is tested electrically using both static and dynamic tests.

### Design of the low parasitic inductance half bridge

The principle for achieving a low parasitic inductance has also been described in [8]. Simply speaking, it's based on having the electrical conducting paths geometry to be overlapped for counteracting the electromagnetic field. Fig. 1 shows the designed half-bridge of the new concept where the power terminals are overlapped. The half-bridge is based on four parallel SiC JFETs and two anti-parallel SiC diodes for each switch (higher and lower switches). These SiC devices are assembled to a stacked structure of direct bonded copper (DBC) substrate, and a flex PCB is used for die-level interconnection. It should be noted here that the anti-parallel diodes are included in the present design due to technology's unclarity level of the SiC devices rated above 1.2kV. With the advance of device technologies making such anti-parallel diodes un-necessary, they can be removed from the present design to further reduce the parasitic inductance.

The stacked DBC substrate (Cu/ceramic/Cu/ceramic/Cu) allows both the power line input and output to be overlapped in the plane directions and forming a parallel plate structure as shown in Fig. 1a&c. This geometry is to cancel the electromagnetic field as much as possible and hence minimize the stray parasitic inductance generated when the current flows through these plates. The connection between the top and middle Cu layers is achieved by Cu filled vias through the upper ceramic layer (Fig. 1a).

The flex PCB is mainly used as interconnection to replace the bond wires and the busbars. It consists of two conductive layers. The lower layer connects the devices to the top conductive

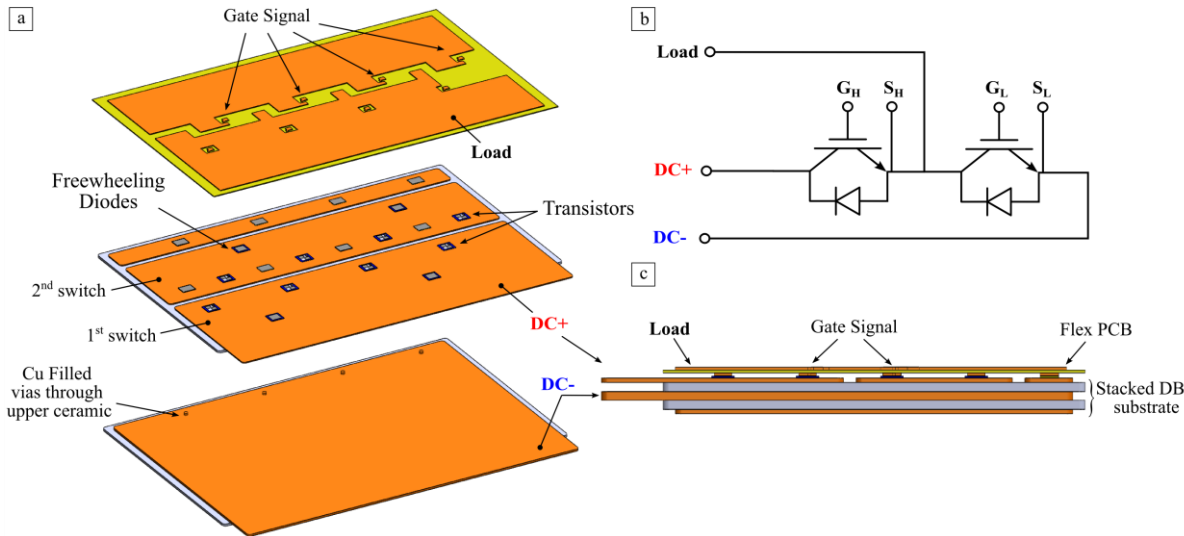


Fig. 1: Strip (a) Exploded and (c) Cross-sectional views of the proposed design showing the stacked DBC structure and Flex PCB on the top, and (b) illustrate the half bridge topology of the proposed design. It should be noted that each switch contains 4 parallel JFETs and

layer of the flex. The latter contains:

- The connection to AC track designed in a way to be overlapped in the plane directions with DC+ track.
- The connection of the source pads of 1<sup>st</sup> switch to drain pads of the 2<sup>nd</sup> switch.
- The connection of source of the 2<sup>nd</sup> leg to DC- track on the DBC.
- The connections for the gate signals to control the JFETs.

It should be noted that the designed package could be compatible with a voltage rate up to 2.5kV for any design margin the device manufacturer takes on the nominal breakdown voltage by tailoring the thickness of insulating materials.

## Electromagnetic simulation of the new package design

The commercial electromagnetic simulation software, MAXWELL 3D, was used to calculate the parasitic parameters associated with the new package design. Fig. 2a shows a cross-sectional drawing illustrating the two major electrical current loops used for the electromagnetic simulation, where loop 1 illustrates the current flowing from DC+ to DC- through the JFETs of the 1<sup>st</sup> switch and the Diodes of the 2<sup>nd</sup> switch, and Loop 2 illustrates the current flowing from DC- to DC+ through the JFETs of the 2<sup>nd</sup> switch and the Diodes of the 1<sup>st</sup> switch. As the thickness of the ceramic in DBC substrates could be different depending on the applications and the electrical insulation abilities of a power module, two thicknesses, i.e. 300 $\mu$ m and 635 $\mu$ m, were considered. Fig. 2b shows the extracted parasitic inductance as a function of frequency for the two loops with the two thicknesses of the ceramic. The parasitic inductance of the major loops is ranging from 1.7 to 2.2nH and from 2.1 to 2.6nH at 1kHz for 300 $\mu$ m and 635 $\mu$ m thick ceramic, respectively. The inductance of the inner loop becomes extremely low since DC+ and DC- tracks are routed on top of each other as shown in Fig. 1c. These value are comparable to the extracted inductance value reported in [9] where they are using FR4 PCB to achieve a similar geometry configuration.

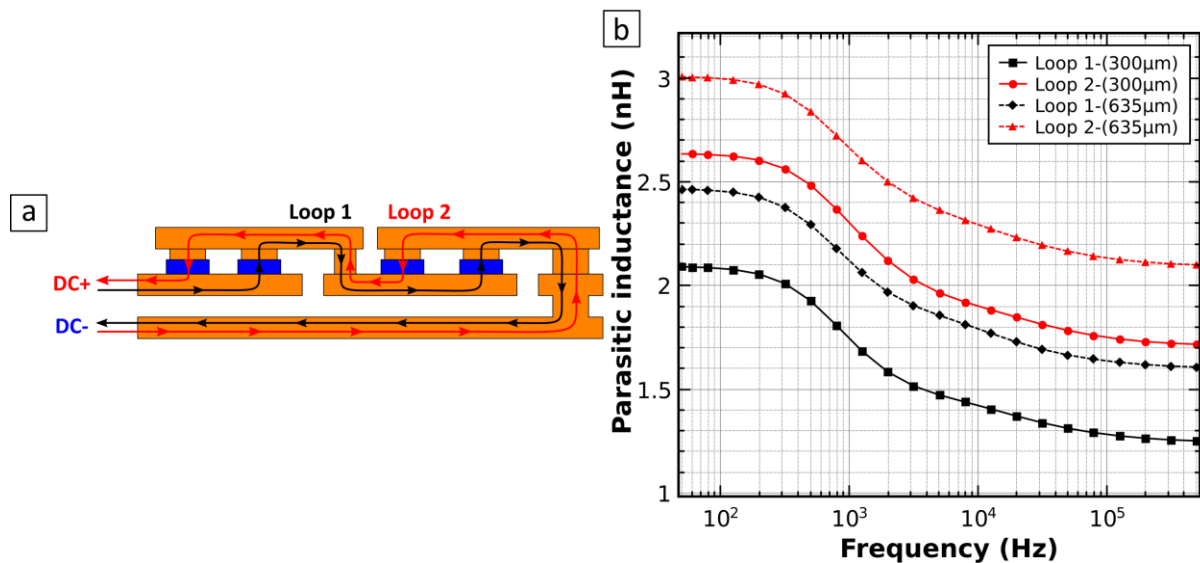


Fig. 2: (a) Illustration of the two electrical current loops used for the electromagnetic simulation used & (b) the quantified inductance of the two loops for two spacing value between the input and output terminal which correspond to the thickness of ceramic layer in the stacked DBC substrate.

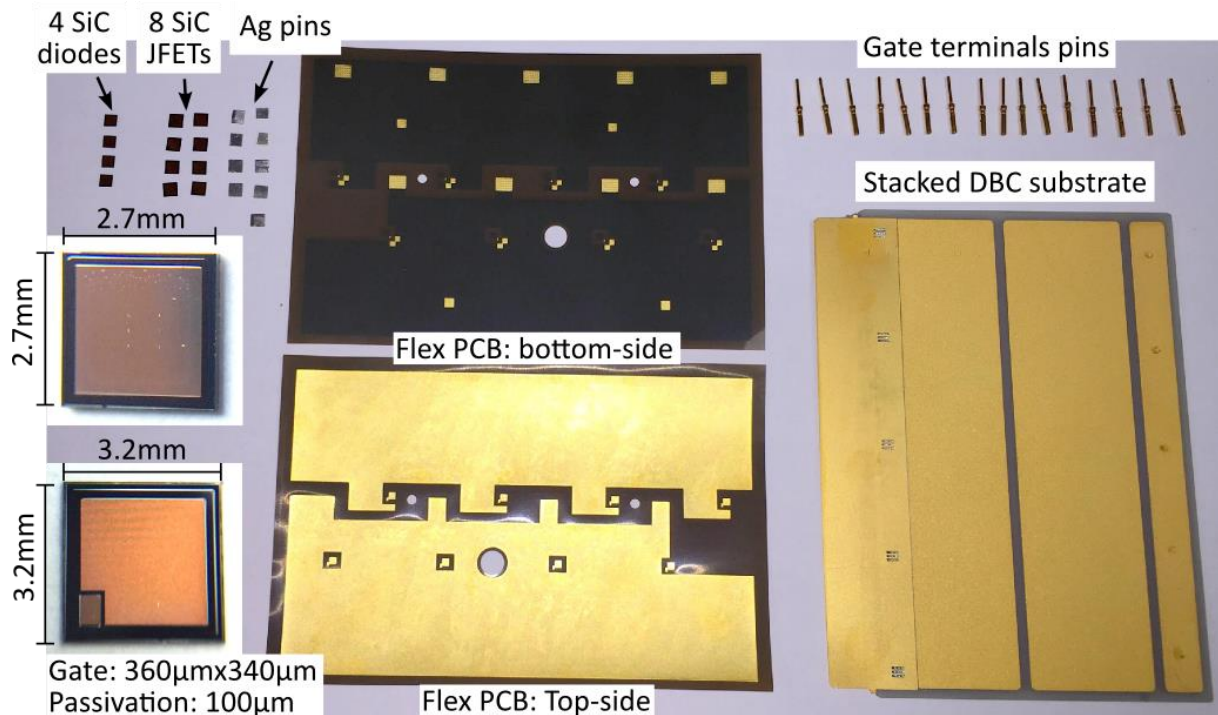


Fig. 3: Parts used for the designed package

## Components and assembling technology

### Parts

Fig. 3 shows the parts used to build the designed package. These are described as follows:

- 4x 1.2kV/2.7mmx2.7mm SiC Diodes with Cu-based finishing metal layer; 2 Diodes per switch
- 8x 1.2kV/3.2mmx3.2mm SiC JFETs with Cu-based finishing metal layer; 4 JFETs per switch
- 635µm thick AlN-based stacked DBC substrate
- Flex PCB
- Pins for Gate terminal connections
- Ag pins for connecting the Flex PCB to the stacked DBC substrates

### Assembly process

Fig. 4 illustrates the presently assembling process flow. Prior to the assembling process, both the Flex PCB and Cu-based SiC JFETs and SiC Diodes are chemically cleaned using acetone and isopropanol solutions. Plasma etching was then used to remove any residual organic contamination and for deoxidization of SiC devices.

First, solder paste was stencil-printed on the copper track of the Flex PCB. Because the 3.2mmx3.2mm SiC JFETs have relatively fine patterns, an accurate die bonder had been employed for accurate die placing to avoid potential short-circuiting between the gate and source. The Flex PCB with the SiC devices was then heated to melt the solder paste and to form the connection between the Cu tracks of the Flex PCB to the Cu track on the topside of the SiC devices. Fig. 4d shows the SiC devices and the Ag pins bonded on the top of the Flex PCB.

Once all the components are soldered to the Flex PCB, the backsides of them will be further bonded to the stacked DBC substrate. Fig. 4e shows the part after that stage with the Gate

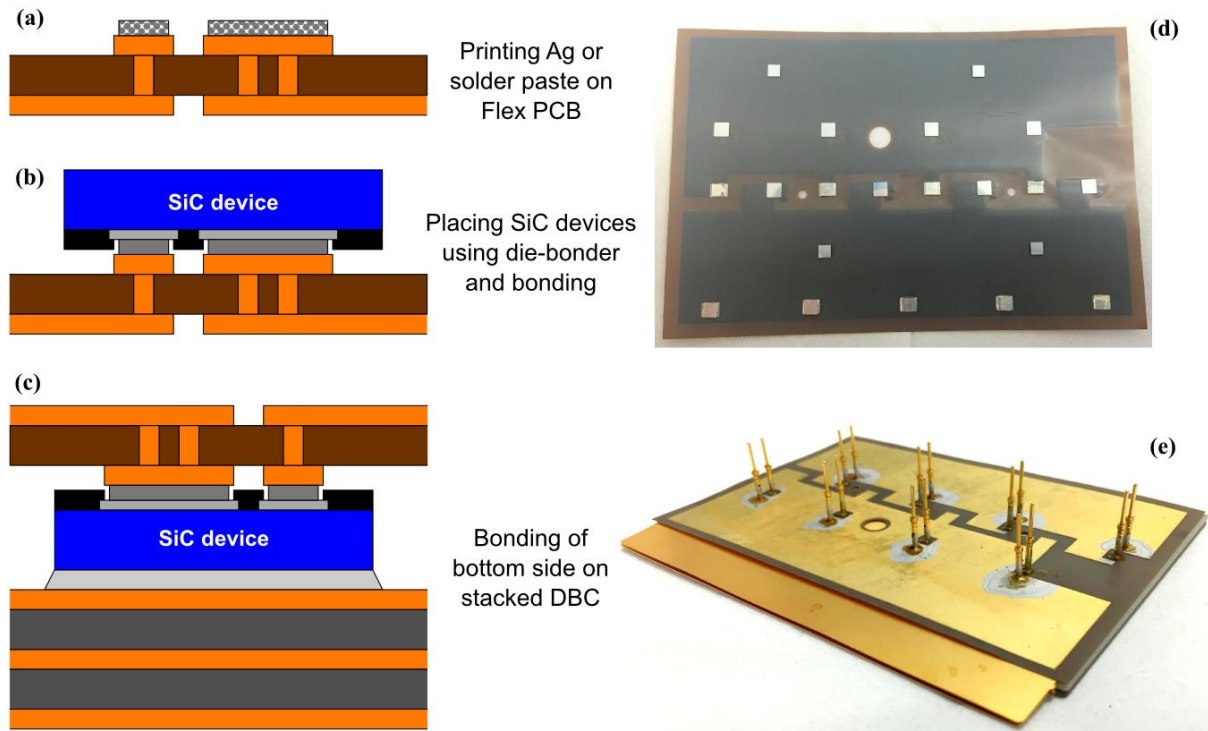


Fig. 4: Illustration of the process flow for assembling the designed package  
 Terminals pins soldered on the topside of the Flex PCB. Finally, insulating silicone gel was injected into the gap between the stacked DBC substrate and the Flex PCB.

## Electrical test

### Static test

A static current-voltage characteristics test was carried out to test the forward characteristics of the JFETs at room temperature. Fig. 5 illustrates the result for 4 parallel SiC JFETs with different gate voltage. It can be seen the currents at different gate voltages are all comparable with the values of 4 times of the values for single SiC JFET [10]. Therefore it can be concluded that all the 4 SiC JFETs have been effectively controlled by the gate signal which

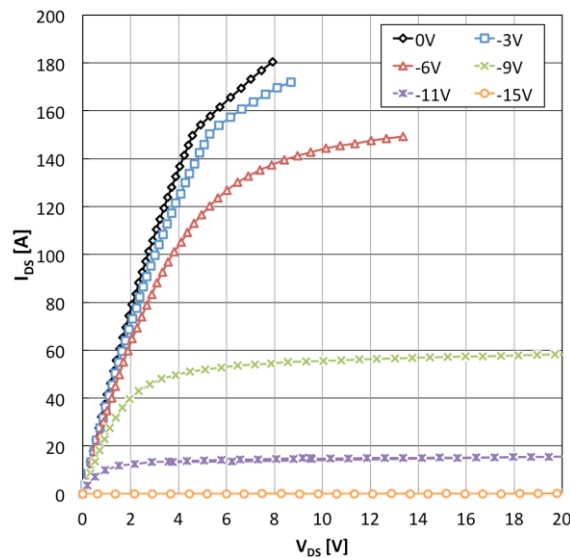


Fig. 5: Forward characteristics of 4 parallel SiC JFETs with different gate voltage

**Table I: Experimental parameters for Double Pulse test**

Part	Type
Drain-Source current ( $I_{DS}$ )	17A
Bus voltage ( $V_{bus}$ )	400V
Gate voltage ( $V_{GS}$ )	0V/-18V
Gate resistance ( $R_g$ )	10 $\Omega$
Load inductance ( $L_{load}$ )	420 $\mu$ H
Junction temperature ( $T_j$ )	25°C (Room temperature)
Oscilloscope	Tektronix DPO 7104 (BW=500MHz)
Voltage probe	Differential probe (BW=50MHz)
Current sensor	Rogowski coil PEM CWT1 (BW=20MHz)

means the designed connection has been achieved.

### Dynamic test

Double-pulse test is a widely used approach to characterize the parasitic electric parameters of power module packaging and to test the switching performance of the latter. It is performed by operating the phase leg (or half bridge) module in a standard switching process including switch-on, conduction, switch-off and block with inductive load, which can control the current going through the switches and voltage applied. This test was carried out at room temperature on the designed package.

The experimental parameters are list below in **Table I**. The lower sides 4 parallel SiC JFETs of the half bridge are the devices under test (DUT) and the inductor is connected in parallel with the higher side switch. The driving signal is switched between 0V and -18V. The load

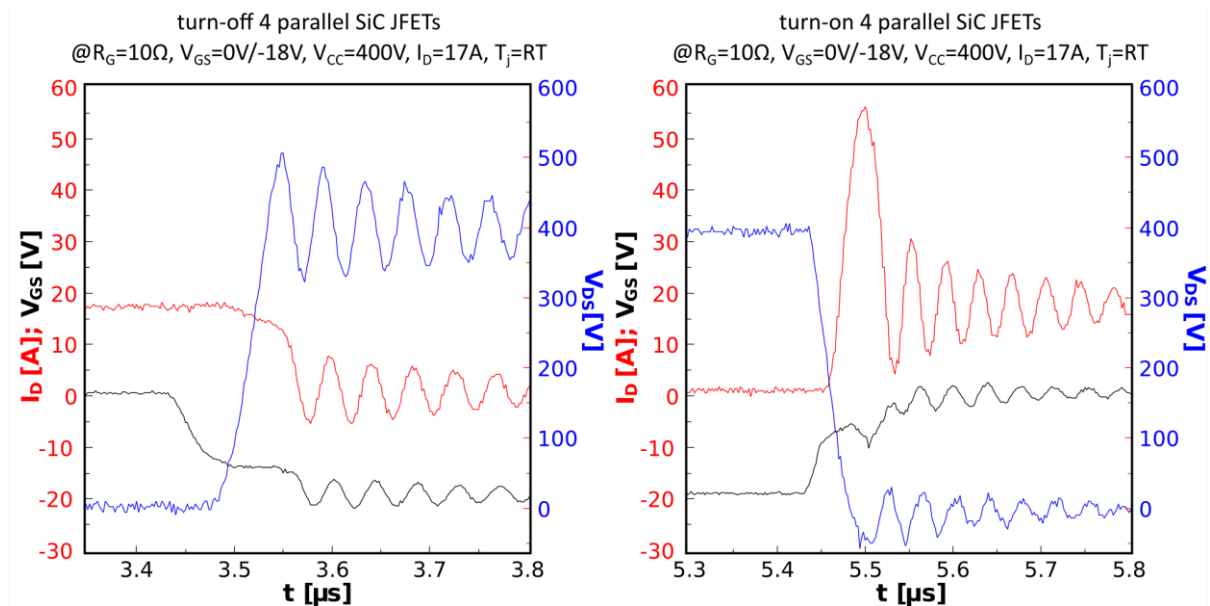


Fig. 6: Double-pulse test at room temperature showing the switching waveforms of 4 parallel SiC JFETs for the turn-off and turn on state ( $V_{gate}=0V/-18V$ ;  $R_g=10\Omega$ )

current is 17A. The DC-Link Voltage is 400V. Fig. 6 presents the voltage and current waveforms of the lower 4x SiC JFETs during the switching off and on transient in the designed package. It can be seen that the voltage overshoot during the turn-off transient is quite low which means the parasitic inductance is low as intended by the design. However, the current overshoot during the turn-on transient is quite high. This is believed to be due to the gate driver as revealed by the gate voltage waveform. It should be mentioned that the gate driver used for this test is an existing gate driver in our lab but not designed specifically for this module.

## Conclusion

With SiC power devices reaching nearly perfect switching properties the parasitic effects around them start to determine the performance. This paper outlined a novel packaging design that reduces drastically these parasitic effects by optimizing the geometry of conductors. An electromagnetic simulation shows a very low parasitic inductance of the DC-link loop. The complete manufacturing process, which offer a high accuracy level to the fine layout of the SiC devices has been described in details. Electrical tests, both static current-voltage characteristics and dynamic transient show a good electrical performance of the 4 parallel SiC JFETs.

## References

- [1] P. Beckedahl, M. Spang, and O. Tamm, "Breakthrough into the third dimension – Sintered multi layer flex for ultra low inductance power modules," in *8th International Conference on Integrated Power Electronics Systems (CIPS)*, 2014, pp. 461–465.
- [2] J. N. Calata, J. G. Bai, X. Liu, S. Wen, and G.-Q. Lu, "Three-Dimensional Packaging for Power Semiconductor Devices and Modules," *IEEE Trans. Adv. Packag.*, vol. 28, no. 3, pp. 404–412, 2005.
- [3] X. He, X. Zeng, X. Yang, and Z. Wang, "A Hybrid Integrated Power Electronic Module Based On Pressure Contact Technology," in *37th IEEE Power Electronics Specialists Conference*, 2006.
- [4] E. Vagnon, J. C. Crebier, Y. Avenas, and P. O. Jeannin, "Study and realization of a low force 3D press-pack power module," in *Power Electronics Specialists Conference, 2008. PESC 2008. IEEE*, 2008, pp. 1048–1054.
- [5] B. Mouawad, M. Soueidan, D. Fabrègue, C. Buttay, B. Allard, V. Bley, H. Morel, and C. Martin, "Application of the Spark Plasma Sintering Technique to Low-Temperature Copper Bonding," *IEEE Trans. Components, Packag. Manuf. Technol.*, vol. 2, no. 4, pp. 553–560, 2012.
- [6] B. Mouawad, B. Thollin, C. Buttay, L. Dupont, V. Bley, D. Fabrègue, M. Soueidan, B. Schlegel, J. Pezard, and J.-C. Crebier, "Direct Copper Bonding for Power Interconnects: Design, Manufacturing and Test," *IEEE Trans. components, Packag. Manuf. Technol.*, vol. 5, no. 1, pp. 143–150, 2015.
- [7] C. M. Johnson, A. Castellazzi, R. Skuriat, P. Evans, J. Li, and P. Agyakwa, "Integrated High Power Modules," in *7th International Conference on Integrated Power Electronics Systems (CIPS)*, 2012, pp. 357–366.
- [8] B. Mouawad, J. Li, A. Castellazzi, C. M. Johnson, T. Erlbacher, and P. Friedrichs, "Low inductance 2.5kV packaging technology for SiC switches," in *9th International Conference on Integrated Power Electronics Systems (CIPS)*, 2016.
- [9] E. Hoene, A. Ostmann, and C. Marczok, "Packaging Very Fast Switching Semiconductors," in *8th International Conference on Integrated Power Electronics Systems (CIPS)*, 2014, pp. 502–508.
- [10] "CoolSiC™ 1200V SiC JFET IJW120R100T1," *Infineon*, 2013. [Online]. Available: [http://www.infineon.com/dgdl/Infineon-IJW120R100T1-DS-v02\\_00-en.pdf?fileId=db3a304341e0aed001420353f03a0e4b](http://www.infineon.com/dgdl/Infineon-IJW120R100T1-DS-v02_00-en.pdf?fileId=db3a304341e0aed001420353f03a0e4b).